

Linear power flow model

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May 17, 2024

Problem setup

Notation

- $j := \sqrt{-1}$
- $I_{n \times n}$ $n \times n$ identity matrix
- 1_n n-dimensional vector of all ones
- $\mathbf{w} := [w_{ij}]_{(i,i) \in \mathcal{E}}$ network parameters; graph edge weights
- $A \in \{0, \pm 1\}^{m \times n}$ edge-to-node incidence matrix
- diag(·) diagonal matrix with the argument as entries
- $\{\cdot\} \circ \{\cdot\}$ elementwise multiplication of two vectors.

Admittance matrix

The admittance matrix $\mathbf{Y} := \mathbf{G} + \mathrm{j} \mathbf{B} \in \mathbb{C}^{n \times n}$ is a generalization of the weighted graph Laplacian matrix. It is a complex, symmetric, but not necessarily Hermitian matrix, and its entries take the form

$$Y_{ik} = \begin{cases} w_{ii} + \sum_{l \neq i} w_{il} & i = k \\ -w_{ik} & \text{otherwise,} \end{cases}$$
 (1)

or equivalently,

$$G_{ik} + jB_{ik} = \begin{cases} \left(g_{ii} + \sum_{l \neq i} g_{il}\right) + j\left(b_{ii} + \sum_{l \neq i} b_{il}\right) & i = k \\ -g_{ik} - jb_{ik} & \text{otherwise.} \end{cases}$$
(2)

Power flow equations

The *power flow equations* are a non-linear system of equations that describe Kirchhoff's current and voltage laws jointly. This system of equations is written as

$$s := p + jq = \operatorname{diag}(u) \underline{Y}\underline{u}, \tag{3}$$

where $\underline{\{\cdot\}}$ is the complex conjugate, $\mathbf{u} \in \mathbb{C}^n$ are the bus voltage phasors and $\mathbf{Y} \in \mathbb{C}^{n \times n}$ is the admittance matrix.

Much ink has been spilled over solving these equations efficiently.

Constructing the power flow manifold

The power flow manifold

Define the *network state* in \mathbb{R}^{4n} as

$$\mathbf{x} := \begin{bmatrix} \mathbf{v}^\mathsf{T} & \boldsymbol{\theta}^\mathsf{T} & \mathbf{p}^\mathsf{T} & \mathbf{q}^\mathsf{T} \end{bmatrix}^\mathsf{T}; \tag{4}$$

the power flow equations define a nonlinear operator $\mathcal{F}: \mathbb{R}^{4n} \to \mathbb{R}^{2n}$, where

$$\mathcal{F}(\mathbf{x}) := \begin{bmatrix} \operatorname{Re} \left\{ \operatorname{diag} \left(\mathbf{u} \right) \underline{Y} \underline{\mathbf{u}} - \mathbf{s} \right\} \\ \operatorname{Im} \left\{ \operatorname{diag} \left(\mathbf{u} \right) \underline{Y} \underline{\mathbf{u}} - \mathbf{s} \right\} \end{bmatrix}$$
 (5)

with $\pmb{u}:=\pmb{v}\circ\exp\left\{\mathrm{j}\pmb{\theta}\right\}$ as the voltage phasors and $\pmb{s}:=\pmb{p}+\mathrm{j}\pmb{q}$ as the complex power injections. The *power flow manifold* is then

$$\mathcal{M} = \left\{ \mathbf{x} \in \mathbb{R}^{4n} : \mathcal{F}(\mathbf{x}) = \mathbf{0}_{2n} \right\}. \tag{6}$$

Linearized power flow manifold

The *linear manifold tangent* to \mathcal{M} at a nominal operating point x_{\bullet} is given by

$$\mathcal{M}_{\bullet} := \left\{ \mathbf{x} \in \mathbb{R}^{4n} : \mathbf{F}(\mathbf{x}_{\bullet}) \left(\mathbf{x} - \mathbf{x}_{\bullet} \right) = \mathbf{0}_{2n} \right\},\tag{7}$$

where

$$F(x_{\bullet}) := \frac{\partial \mathcal{F}}{\partial x}(x_{\bullet}) = \begin{bmatrix} \frac{\partial \mathcal{F}}{\partial v}(x_{\bullet}) & \frac{\partial \mathcal{F}}{\partial \theta}(x_{\bullet}) & \frac{\partial \mathcal{F}}{\partial p}(x_{\bullet}) & \frac{\partial \mathcal{F}}{\partial q}(x_{\bullet}) \end{bmatrix}$$
(8a)
$$= \begin{bmatrix} \operatorname{Re}\left\{\frac{\partial s}{\partial v}(u_{\bullet})\right\} & \operatorname{Re}\left\{\frac{\partial s}{\partial \theta}(u_{\bullet})\right\} & -I_{n \times n} & \mathbf{0}_{n \times n} \\ \operatorname{Im}\left\{\frac{\partial s}{\partial v}(u_{\bullet})\right\} & \operatorname{Im}\left\{\frac{\partial s}{\partial \theta}(u_{\bullet})\right\} & \mathbf{0}_{n \times n} & -I_{n \times n} \end{bmatrix}$$
(8b)
$$= \begin{bmatrix} \frac{\partial p}{\partial v}(u_{\bullet}) & \frac{\partial p}{\partial \theta}(u_{\bullet}) & -I_{n \times n} & \mathbf{0}_{n \times n} \\ \frac{\partial q}{\partial v}(u_{\bullet}) & \frac{\partial q}{\partial \theta}(u_{\bullet}) & \mathbf{0}_{n \times n} & -I_{n \times n} \end{bmatrix} .$$
(8c)

Flat start linearization

Linearization of the power flow manifold [1, 2, 3]

Consider the flat start condition $u_{\star} := 1 + \mathrm{j} \mathbf{0}$, and suppose that $\omega = \mathbf{0} + \mathrm{j} \mathbf{0}$.

Then, the linear power flow manifold around u_{\star} is

$$\mathcal{M}_{\star} := \left\{ x \in \mathbb{R}^{4n} : F(x_{\star})(x - x_{\star}) = \mathbf{0}_{2n} \right\}, \tag{9}$$

where

$$F(x_{\star}) = \begin{vmatrix} G & -B & -I_{n \times n} & \mathbf{0}_{n \times n} \\ -B & -G & \mathbf{0}_{n \times n} & -I_{n \times n} \end{vmatrix}, \tag{10}$$

or equivalently,

$$p = G\epsilon - B\theta$$
, and $q = -B\epsilon - G\theta$, (11)

where $\epsilon := v - 1$.

Flat start linearization, part 1

Let $\omega := \gamma + \mathrm{j}\beta \in \mathbb{C}^n$ denote the vector of self-admittances of each node. Then, following [5, 5.10]

$$\begin{split} \frac{\partial \mathbf{s}}{\partial \boldsymbol{\theta}} \left(\mathbf{u}_{\star} \right) &= \mathrm{j} \, \mathrm{diag} \left(\mathbf{u}_{\star} \right) \left(\mathrm{diag} \left(\underline{Y} \mathbf{u}_{\star} \right) - \underline{Y} \, \mathrm{diag} \left(\underline{\mathbf{u}}_{\star} \right) \right) \\ &= \mathrm{j} \mathbf{I}_{n \times n} \left(\mathrm{diag} \left(\underline{Y} \mathbf{1}_{n} \right) - \underline{Y} \mathbf{I}_{n \times n} \right) \\ &= \mathrm{j} \left(\mathrm{diag} \left(\underline{\boldsymbol{\omega}} \right) - \underline{Y} \right) \end{split}$$

and

$$\begin{split} \frac{\partial \mathbf{s}}{\partial \mathbf{v}}(\mathbf{u}_{\star}) &= \operatorname{diag}(\mathbf{u}) \left(\operatorname{diag}(\underline{Y}\underline{\mathbf{u}}) + \underline{Y} \operatorname{diag}(\underline{\mathbf{u}}) \right) \operatorname{diag}(\mathbf{v})^{-1} \\ &= I_{n \times n} \left(\operatorname{diag}(\underline{Y}\mathbf{1}_n) + \underline{Y}\mathbf{I}_{n \times n} \right) I_{n \times n}^{-1} \\ &= \operatorname{diag}(\underline{\omega}) + \underline{Y} \end{split}$$

Flat start linearization, part 2

We have that

$$\begin{split} \frac{\partial \mathbf{p}}{\partial \boldsymbol{\theta}} \left(\mathbf{u}_{\star} \right) &:= \operatorname{Re} \left\{ \frac{\partial \mathbf{s}}{\partial \boldsymbol{\theta}} \left(\mathbf{u}_{\star} \right) \right\} \\ &= \operatorname{Re} \left\{ \operatorname{j} \left(\operatorname{diag} \left(\underline{\omega} \right) - \underline{Y} \right) \right\} \\ &= \operatorname{Re} \left\{ \operatorname{j} \left(\operatorname{diag} \left(\gamma - \operatorname{j} \boldsymbol{\beta} \right) - \left(\boldsymbol{G} - \operatorname{j} \boldsymbol{B} \right) \right) \right\} \\ &= \operatorname{Re} \left\{ \operatorname{j} \left(\operatorname{diag} \left(\gamma \right) + \operatorname{diag} (\boldsymbol{\beta}) - \operatorname{j} \boldsymbol{G} - \boldsymbol{B} \right\} \right\} \\ &= \operatorname{diag} \left(\boldsymbol{\beta} \right) - \boldsymbol{B}, \end{split}$$

and

$$egin{aligned} rac{\partial oldsymbol{p}}{\partial oldsymbol{v}} \left(oldsymbol{u}_{\star}
ight) &:= \mathsf{Re} \left\{ \mathsf{diag} \left(oldsymbol{\omega}
ight) + oldsymbol{Y} - \mathrm{j} oldsymbol{\mathcal{B}}
ight\} \ &= \mathsf{diag} \left(\gamma
ight) + oldsymbol{G}. \end{aligned}$$

Flat start linearization, part 3

Similarly,

$$egin{aligned} rac{\partial \mathbf{q}}{\partial heta} \left(\mathbf{u}_{\star}
ight) &= \operatorname{Im} \left\{ rac{\partial \mathbf{s}}{\partial heta} \left(\mathbf{u}_{\star}
ight)
ight\} \ &= \operatorname{Im} \left\{ \operatorname{j} \left(\operatorname{diag} \left(\underline{\omega}
ight) - \underline{Y}
ight)
ight\} \ &= \operatorname{Im} \left\{ \operatorname{j} \operatorname{diag} (\gamma) + \operatorname{diag} (eta) - \operatorname{j} oldsymbol{G} - oldsymbol{B}
ight\} \ &= \operatorname{diag} \left(\gamma
ight) - oldsymbol{G} \end{aligned}$$

and

$$egin{aligned} & rac{\partial oldsymbol{q}}{\partial oldsymbol{v}} \left(oldsymbol{u}_{\star}
ight) = \operatorname{Im} \left\{ \operatorname{diag} \left(oldsymbol{\omega}
ight) + oldsymbol{Y}
ight\} \ &= -\operatorname{diag} \left(oldsymbol{eta}
ight) - oldsymbol{B} \end{aligned}$$

Applying the assumption that $\beta = \gamma = 0$ and plugging into (8) yields the desired result:

$$F(x_{\star}) = \begin{bmatrix} G & -B & -I_{n \times n} & 0_{n \times n} \\ -B & -G & 0_{n \times n} & -I_{n \times n} \end{bmatrix}.$$
 (12)

Inverse model for tree networks

Remark: Model inversion

In the special case of *radial (tree) networks*, i.e., the practically relevant setting of small-scale *distribution networks* (the main application setting of interest), a reduced form of the linear model (11) can be inverted in closed form.

Trees only

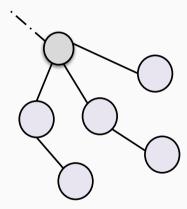


Figure 1: Hereafter, everything only works for tree networks

Reduced incidence matrix

With an abuse of notation, for a *tree network* with n lines and n non-reference nodes, $\mathcal{N} := \{0,1,\dots,n\}$, let $\mathbf{A} \in \{0,\pm 1\}^{n\times n}$ be the reduced incidence matrix formed by removing the first column of the full incidence matrix, which we now denote as

$$\mathbf{A}_{\star} := \begin{bmatrix} \mathbf{a}_0 & \mathbf{A} \end{bmatrix} \in \{0, \pm 1\}^{n \times (n+1)}. \tag{13}$$

It is known that the reduced incidence matrix **A** is square and invertible [4].

Reduced Laplacian matrices

With another abuse of notation, keep the same formula for the admittance matrix Y,

$$Y := A^{\mathsf{T}} \operatorname{diag}(w) A = A^{\mathsf{T}} \operatorname{diag}(g) A + j A^{\mathsf{T}} \operatorname{diag}(b) A, \tag{14}$$

and define the following reduced graph Laplacian matrices:

$$G := \operatorname{Re} \{Y\} = A^{\mathsf{T}} \operatorname{diag}(g)A$$
, and $B := \operatorname{Im} \{Y\} = A^{\mathsf{T}} \operatorname{diag}(b)A$, (15)

where **A** is now the reduced incidence matrix.

Now the inverse of Y, the impedance matrix, can be computed as

$$Y^{-1} = A^{-1} \operatorname{diag}(w)^{-1} A^{-T} = R + jX.$$
 (16)

Intuitive explanation

Recall that for any complex number ζ , $\zeta^{-1} = \underline{\zeta}/|\zeta|^2$. Hence, the impedances can be expressed as inverse admittances (graph weights):

$$w_{ij}^{-1} := r_{ij} + jx_{ij} := \underbrace{\frac{g_{ij}}{g_{ij}^2 + b_{ij}^2}}_{:=r_{ii}} + j\underbrace{\frac{-b_{ij}}{g_{ij}^2 + b_{ij}^2}}_{:=x_{ii}} \quad \forall (i,j) \in \mathcal{E},$$
(17)

thus, define the matrices

$$R = A^{-1}\operatorname{diag}(r)A^{-T}, \quad X := A^{-1}\operatorname{diag}(x)A^{-T}.$$

Inverse model for tree networks

Inverse model for tree networks

Consider a tree network with n non-reference nodes and n edges. Then,

$$\begin{bmatrix} G & -B \\ -B & -G \end{bmatrix}^{-1} = \begin{bmatrix} R & X \\ X & -R \end{bmatrix},$$

where

$$\mathbf{R} = \mathbf{A}^{-1} \operatorname{diag}(\mathbf{r}) \mathbf{A}^{-T}$$
,

$$X := A^{-1} \operatorname{diag}(x) A^{-T}.$$

and thus

$$ig| \epsilon = R p + X q$$
, and $heta = X p - R q$,

where $\epsilon := \mathbf{v} - \mathbf{1}$.

(19)

(18)

Inverting the model for trees

We want to invert the system of equations

$$\begin{bmatrix} \mathbf{p} \\ \mathbf{q} \end{bmatrix} = \begin{bmatrix} \mathbf{G} & -\mathbf{B} \\ -\mathbf{B} & -\mathbf{G} \end{bmatrix} \begin{bmatrix} \boldsymbol{\epsilon} \\ \boldsymbol{\theta} \end{bmatrix} .$$
 (20)

Note that as $G \succ 0$, both Schur complements of the above matrix exist, and hence

$$\begin{bmatrix} G & -B \\ -B & -G \end{bmatrix}^{-1} = \begin{bmatrix} \left(G + BG^{-1}B\right)^{-1} & \mathbf{0}_{n \times n} \\ \mathbf{0}_{n \times n} & -\left(G + BG^{-1}B\right)^{-1} \end{bmatrix} \begin{bmatrix} \mathbf{I}_{n \times n} & -BG^{-1} \\ BG^{-1} & \mathbf{I}_{n \times n} \end{bmatrix}$$
$$= \begin{bmatrix} \left(G + BG^{-1}B\right)^{-1} & -\left(G + BG^{-1}B\right)^{-1}BG^{-1} \\ -\left(G + BG^{-1}B\right)^{-1}BG^{-1} & -\left(G + BG^{-1}B\right)^{-1} \end{bmatrix}.$$

Inverting the model for trees

To begin, the first matrix we need to compute is

$$\left(\mathbf{G} + \mathbf{B} \mathbf{G}^{-1} \mathbf{B} \right)^{-1} = \left(\mathbf{A}^{\mathsf{T}} \operatorname{diag}(\mathbf{g}) \mathbf{A} + \mathbf{A}^{\mathsf{T}} \operatorname{diag}(\mathbf{b}) \mathbf{A} \mathbf{A}^{-1} \operatorname{diag}(\mathbf{g})^{-1} \mathbf{A}^{-\mathsf{T}} \mathbf{A}^{\mathsf{T}} \operatorname{diag}(\mathbf{b}) \mathbf{A} \right)^{-1}$$

$$= \left(\mathbf{A}^{\mathsf{T}} \operatorname{diag} \left(\left[\frac{\mathbf{g}_{ij}^2 + b_{ij}^2}{\mathbf{g}_{ij}^2 + b_{ij}^2} \right]_{ij \in \mathcal{E}} \right) \mathbf{A} \right)^{-1}$$

$$= \mathbf{A}^{-1} \operatorname{diag} \left(\left[\frac{\mathbf{g}_{ij}}{\mathbf{g}_{ij}^2 + b_{ij}^2} \right]_{ij \in \mathcal{E}} \right) \mathbf{A}^{-\mathsf{T}}$$

$$:= \mathbf{A}^{-1} \operatorname{diag}(\mathbf{r}) \mathbf{A}^{-\mathsf{T}}$$

$$:= \mathbf{R}.$$

Inverting the model for trees

Finally, the second matrix we need to compute is

$$-\left(G + BG^{-1}B\right)^{-1}BG^{-1} = -A^{-1}\operatorname{diag}\left(\left[\frac{g_{ij}}{g_{ij}^2 + b_{ij}^2}\right]_{ij \in \mathcal{E}}\right)A^{-T}A^{T}\operatorname{diag}(b)AA^{-1}\operatorname{diag}(g)^{-1}A^{-T}$$

$$= A^{-1}\operatorname{diag}\left(\left[\frac{-b_{ij}}{g_{ij}^2 + b_{ij}^2}\right]_{ij \in \mathcal{E}}\right)A^{-T}$$

$$:= A^{-1}\operatorname{diag}(x)A^{-T}$$

$$:= X$$

The matrices of the inverted model

Then, the matrix inverse we have obtained is

$$\begin{bmatrix} G & -B \\ -B & -G \end{bmatrix}^{-1} = \begin{bmatrix} R & X \\ X & -R \end{bmatrix}$$

and thus

$$\begin{bmatrix} \epsilon \\ \theta \end{bmatrix} = \begin{bmatrix} R & X \\ X & -R \end{bmatrix} \begin{bmatrix} p \\ q \end{bmatrix},$$

as desired.

(22)

(21)

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